

Novel Measurement-Noise-Suppression and Measurement-Time-Reduction Methodology for ADC/DAC

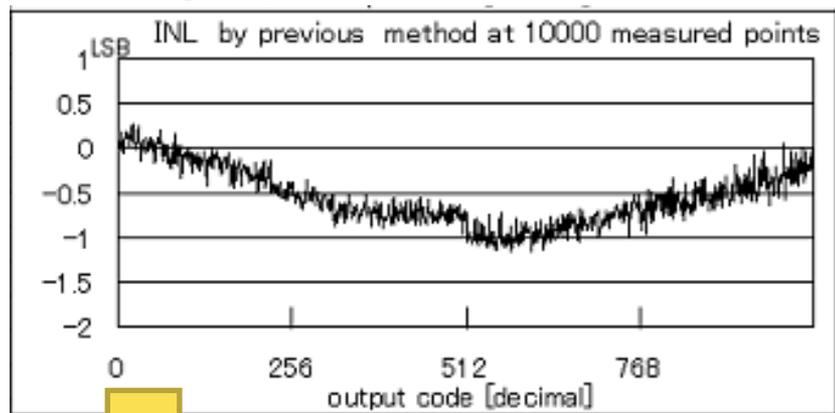
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菅原 光俊^{*+}、松澤 昭^{*}

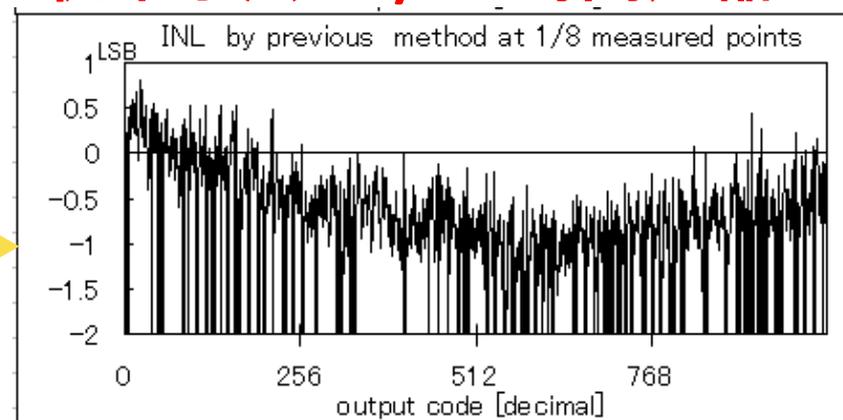
Tokyo Institute of Technology^{*} & Consultant⁺

Mar 10, 2014

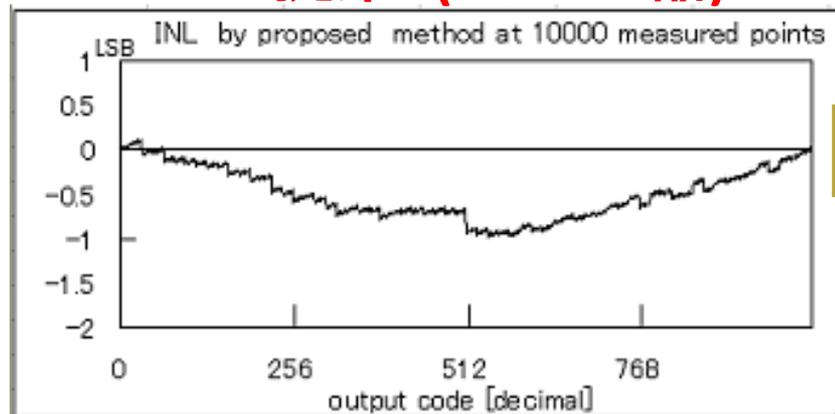
従来



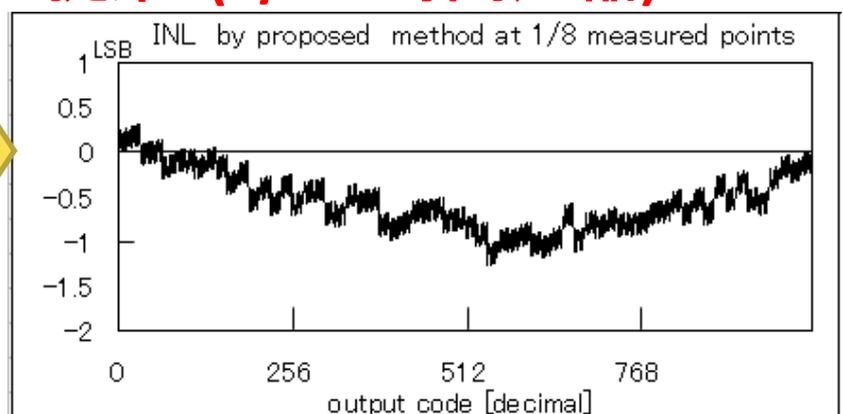
従来手法で1/8に時間短縮



提案1(ノイズ圧縮)

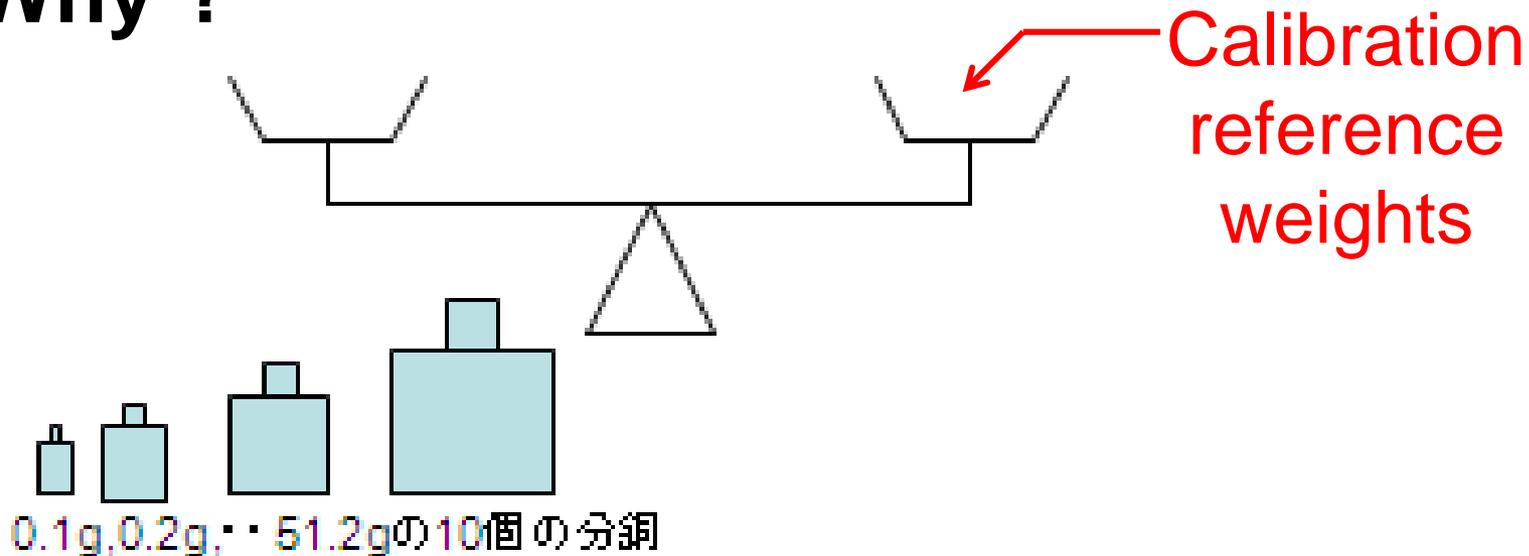


提案2(1/8に時間短縮)



- Previously Histogram-method or Moment method are used for ADC, DAC measurement.
- I am proposing a novel method to **reduce noise to 1/22** at 10bit binary ADC. Or to **reduce measurement time 1/8**, when previous noise level is acceptable.
 - First, calculate each physical weight value statistically.
 - Next, reproduce each step using the weights.

- On 2pan-balance scale certification, only need to accurately measure each weights.
- On test method of ADC, measured calibration reference weights 0.01g, 0.02g , - - , 102.2g, 102.3g, total 10240 times !
Why ?



Why ADC is measured $>10,000$ times ?

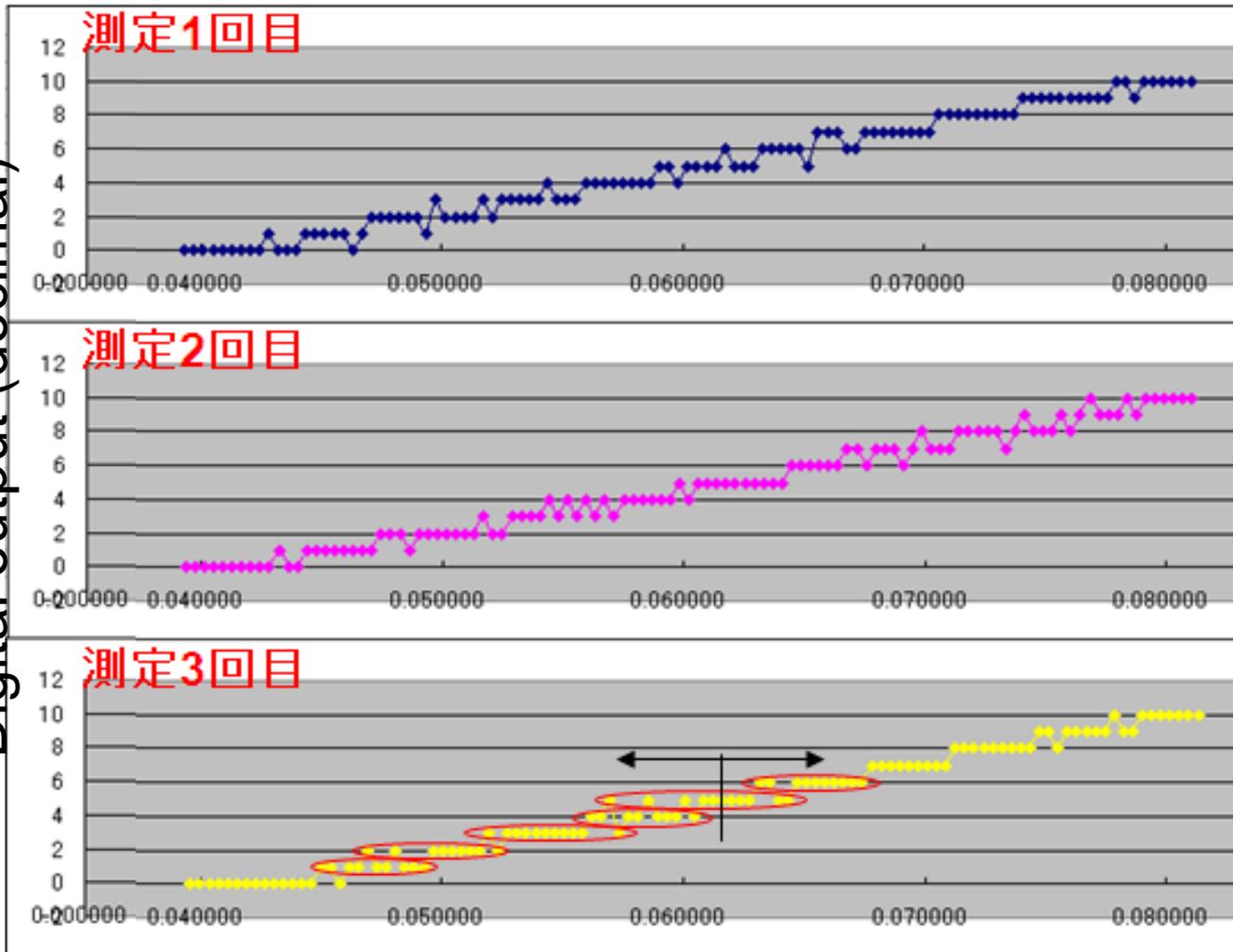
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- **“ADC cannot be measured each internal weight directly, thus it is necessary to sweep inputs to find transition points.”**
 - To measure 1.6g weight under test, reference weights will be 1.55g, 1.56g, ..., 1.64g, 1.65g, to find balance points.
- **“ADC has more than 1LSB noise, therefore need to reduce noise.”**
 - Needle fluctuation, or stop at various positions time to time. Need averaging.

Previous Histogram/Moment method / 6

Digital output (decimal)



Analog input voltage

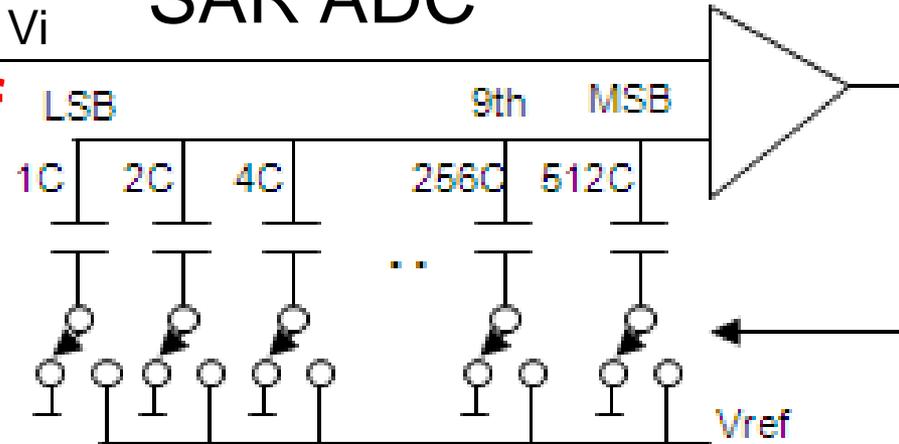
- Sweep input voltage by ramp
- Outputs fluctuate by noise
- Histogram method ignores x-axis positions - always monotonic
- Moment method uses averaged x per y value

A case of 10bit **binary coded** ADC

Proposed method for binary weighted 1 / 8

- Start from same ramp data;

• Calculate 512C weight as V_i SAR ADC differences of MSB on/off



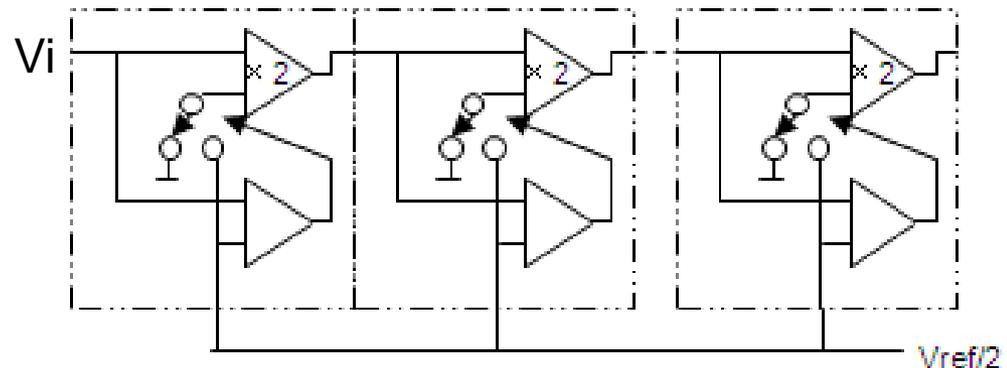
- $V_i(512C) = V_i(512) - V_i(0)$
- $V_i(512C) = V_i(513) - V_i(1)$
- $V_i(512C) = V_i(514) - V_i(2)$
- $V_i(512C) = V_i(1023) - V_i(511)$
- $V_i(1x \text{ xxxx } xxxx)$
- $V_i(0x \text{ xxxx } xxxx)$

- Average above;

- $\overline{V_i(512C)}$
- Noise sigma becomes $1/\sqrt{512} = 1/22$.

Noise reduced 1/22 !

Pipelined ADC



- Calculate 9th bit weight.

- $V_i(256C) = V_i(256) - V_i(0)$

- $V_i(256C) = V_i(257) - V_i(1)$

:

- $V_i(256C) = V_i(511) - V_i(255)$

- $V_i(256C) = V_i(768) - V_i(512)$

- $V_i(256C) = V_i(769) - V_i(513)$

:

- $V_i(256C) = V_i(1023) - V_i(767)$

$$\begin{array}{r} V_i(x1 \text{ xxxx xxxx}) \\ -) \underline{V_i(x0 \text{ xxxx xxxx})} \end{array}$$

- Average above;

- $\overline{V_i(256C)}$

- Noise sigma becomes $1/\sqrt{512} = 1/22$.

- Get each noise-suppressed weight

- $\overline{V_i(512C)}, \overline{V_i(256C)}, \dots, \overline{V_i(1C)}$

- Each noise sigma becomes $1/\sqrt{512} = 1/22$.

- **Reproduce noise-reduced ADC**

- $V_{re}(0)=0$

- $V_{re}(1)=\overline{V_i(\text{LSB})}$

- $V_{re}(2)=\overline{V_i(2^{\text{nd}})}$

- $V_{re}(3)=\overline{V_i(2^{\text{nd}})+V_i(\text{LSB})}$

⋮

- $V_{re}(513)=\overline{V_i(\text{MSB})+V_i(\text{LSB})}$

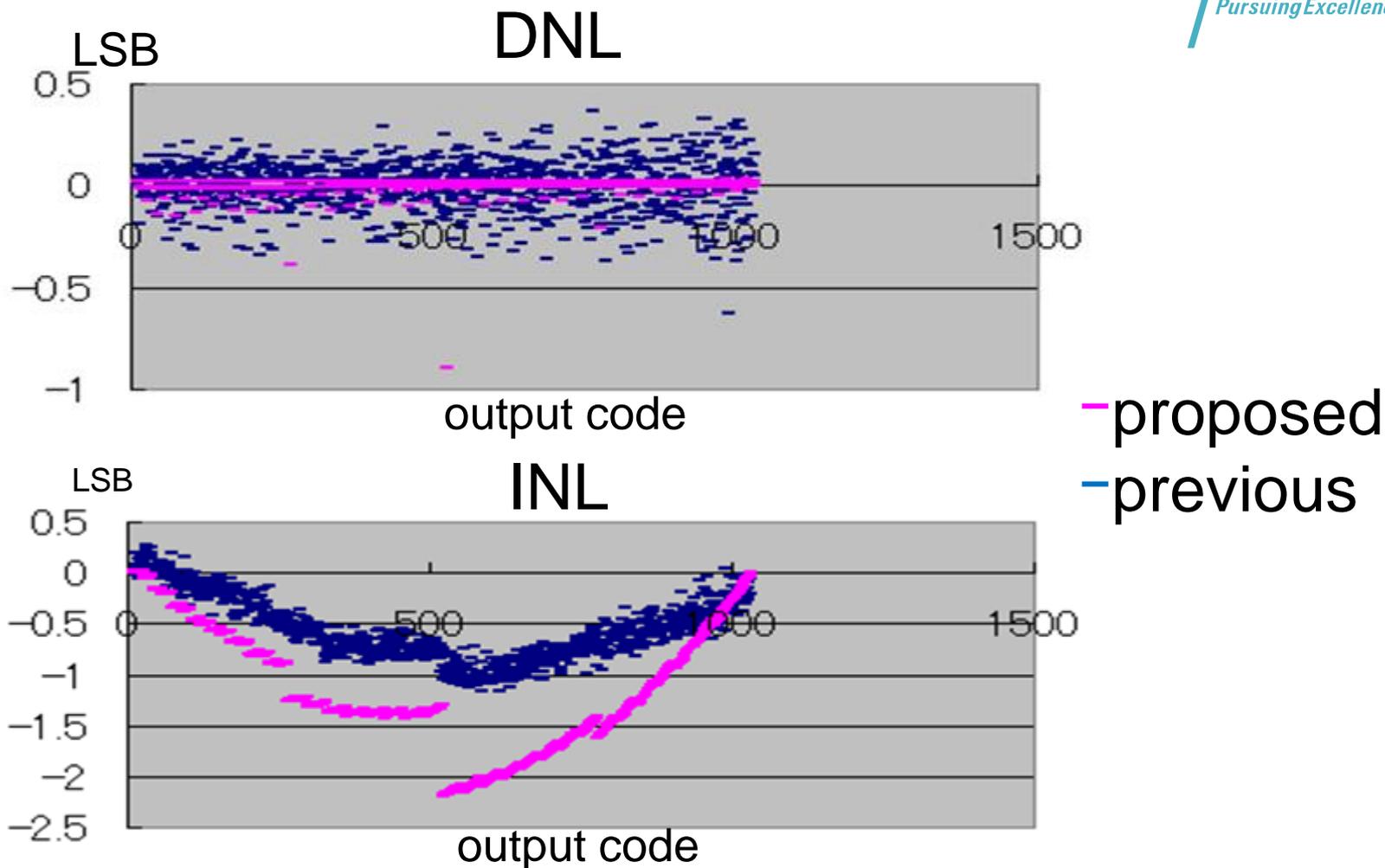
⋮

- $V_{re}(1023)=\overline{V_i(\text{MSB})+V_i(9^{\text{th}})+V_i(8^{\text{th}})+\dots+V_i(\text{LSB})}$

- Calculate linearity

- INL

- DNL



I have detected that DUT was **not binary coded !**

**A case of 10bit ADC consists of
thermometer coded at top 6bits
+binary coded at lower 4bits**

- Calculate binary weight at lower 4bits

- $V_i(8C) = V_i(8) - V_i(0)$
 - $V_i(8C) = V_i(9) - V_i(1)$
 - \vdots
 - $V_i(8C) = V_i(15) - V_i(7)$
 - $V_i(8C) = V_i(24) - V_i(16)$
 - $V_i(8C) = V_i(25) - V_i(17)$
 - \vdots
 - $V_i(8C) = V_i(1016) - V_i(1008)$
 - $V_i(8C) = V_i(1023) - V_i(1015)$
- $V_i(\text{xx xxxx } \mathbf{1} \text{xxx})$
 -) $V_i(\text{xx xxxx } \mathbf{0} \text{xxx})$

- Average above;
 - $\overline{V_i(8C)}$
 - Noise sigma becomes $1/\sqrt{512} = 1/22$.
- Get each noise-reduced weight
 - $\overline{V_i(8C)}, \overline{V_i(4C)}, \overline{V_i(2C)}, \overline{V_i(1C)}$
 - Each noise sigma becomes $1/\sqrt{512} = 1/22$.

- **Reproduce noise-reduced ADC**

- $V_{re}(0)=0$
- $V_{re}(1)=\overline{V_i(1C)}$
- $V_{re}(2)=\overline{V_i(2C)}$
- $V_{re}(3)=\overline{V_i(1C)}+\overline{V_i(2C)}$
- \vdots
- $V_{re}(16)=\overline{V_i(16C_1)}$
- $V_{re}(16)=\overline{V_i(16C_1)}+\overline{V_i(1C)}$
- \vdots
- $V_{re}(33)=\overline{V_i(16C_1)}+\overline{V_i(16C_2)}+\overline{V_i(1C)}$
- \vdots
- $V_{re}(1023)=\overline{V_i(16C_1)}+\overline{V_i(16C_2)}+\dots+\overline{V_i(8C)}+\dots+\overline{V_i(1C)}$

- Calculate linearity

- INL
- DNL

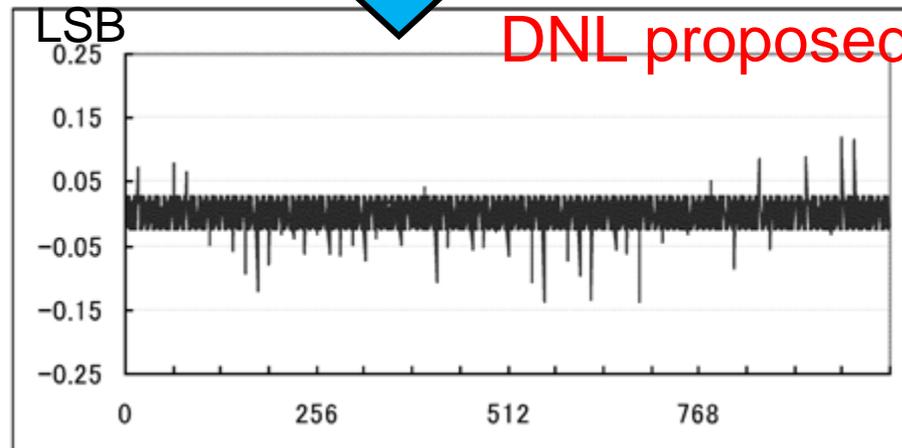
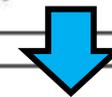
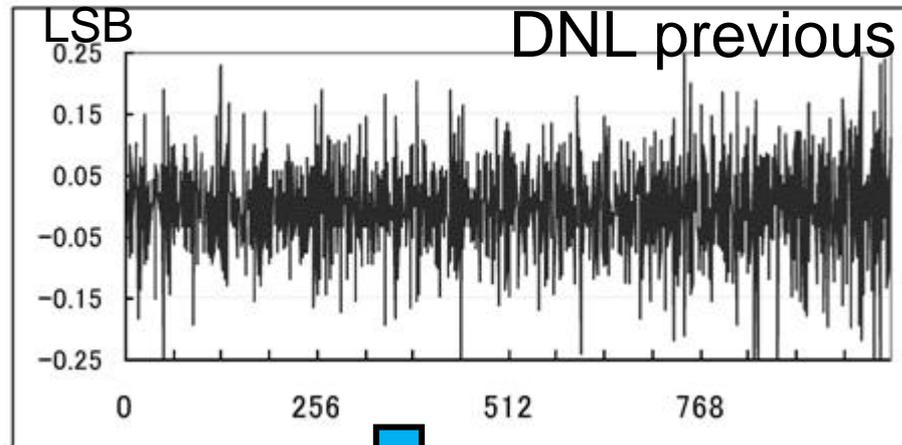
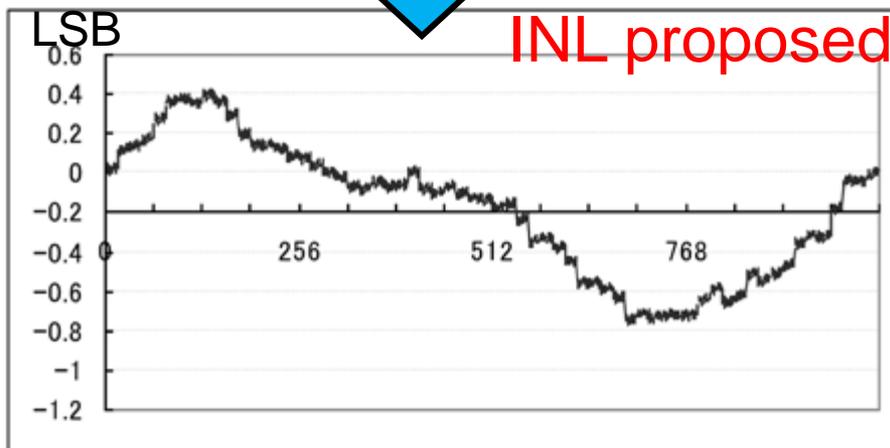
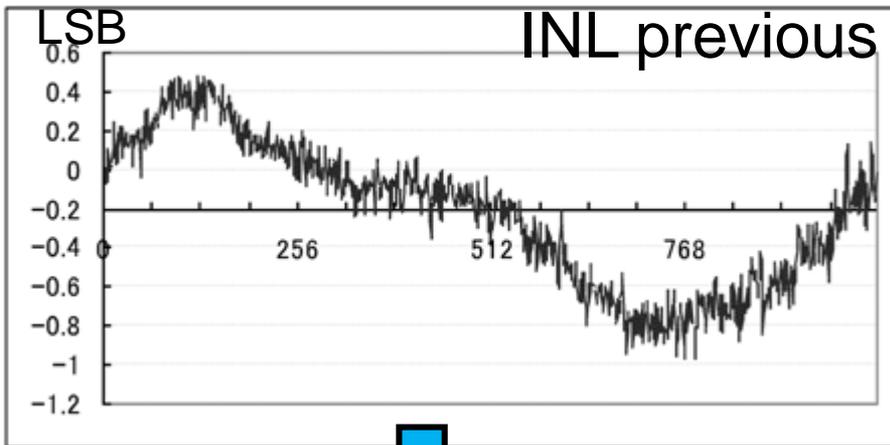
Experiment results 1

Noise reduction

Noise becomes 1/4 ~ 1/22

10bit ADC,

10240 measuring points



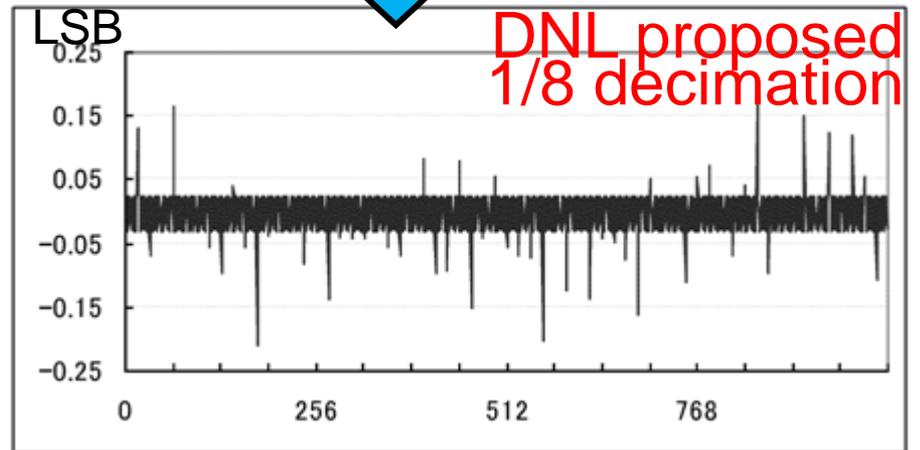
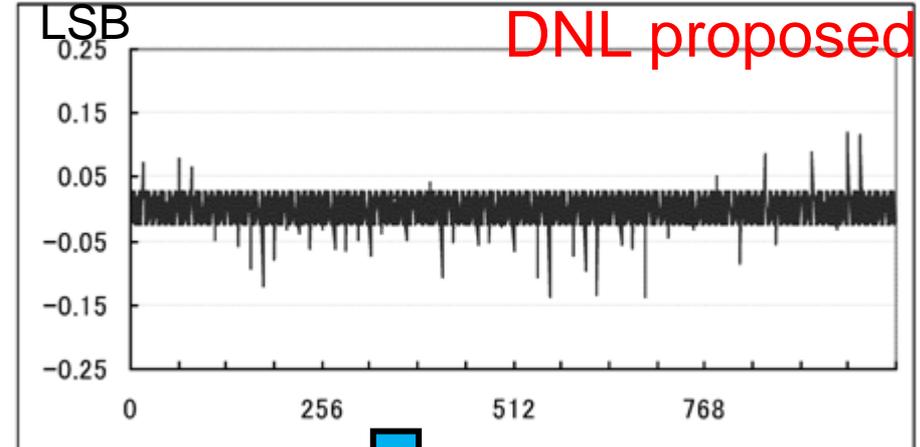
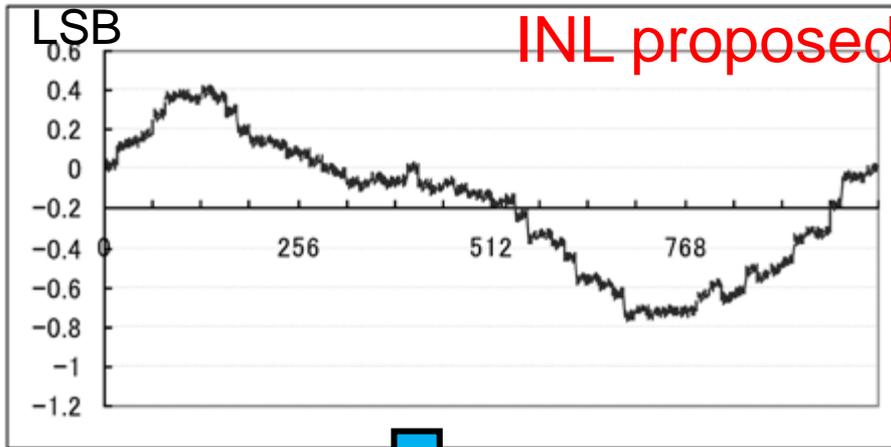
(Equivalence of x16 = 160,000 measuring points or more)

Novel Measuring-Noise-Suppression and
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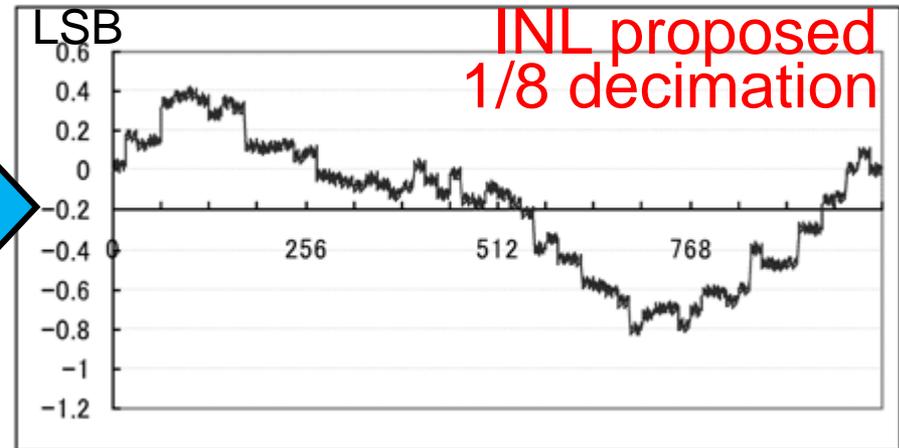
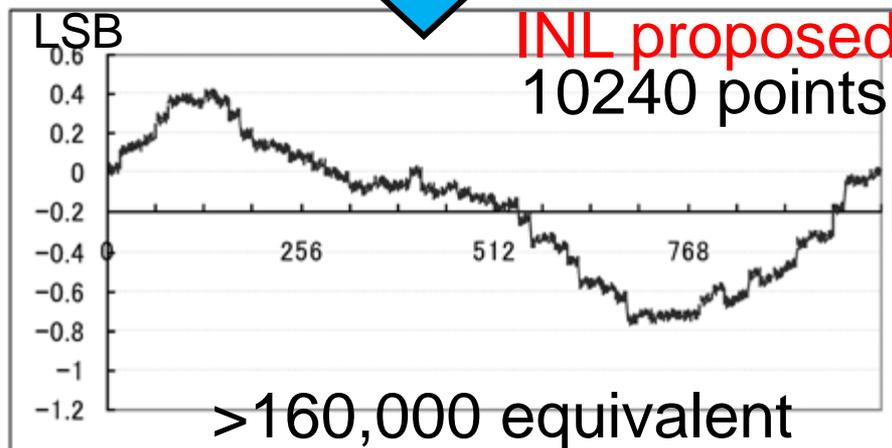
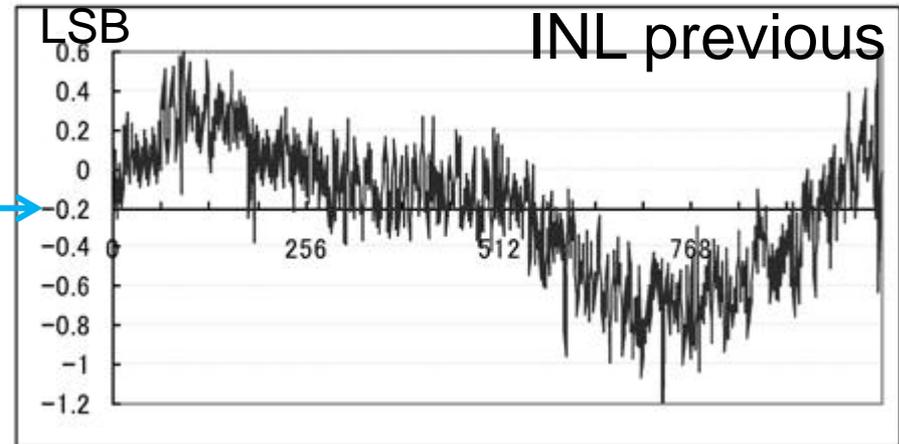
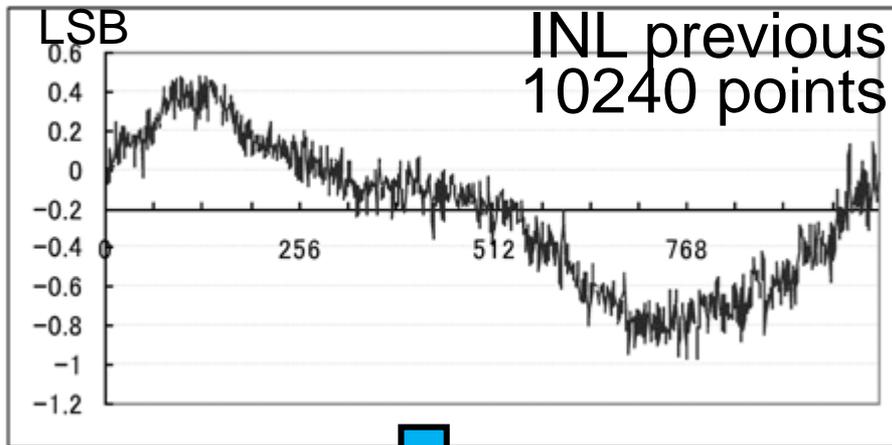
Experiment results 2

Measurement time reduction

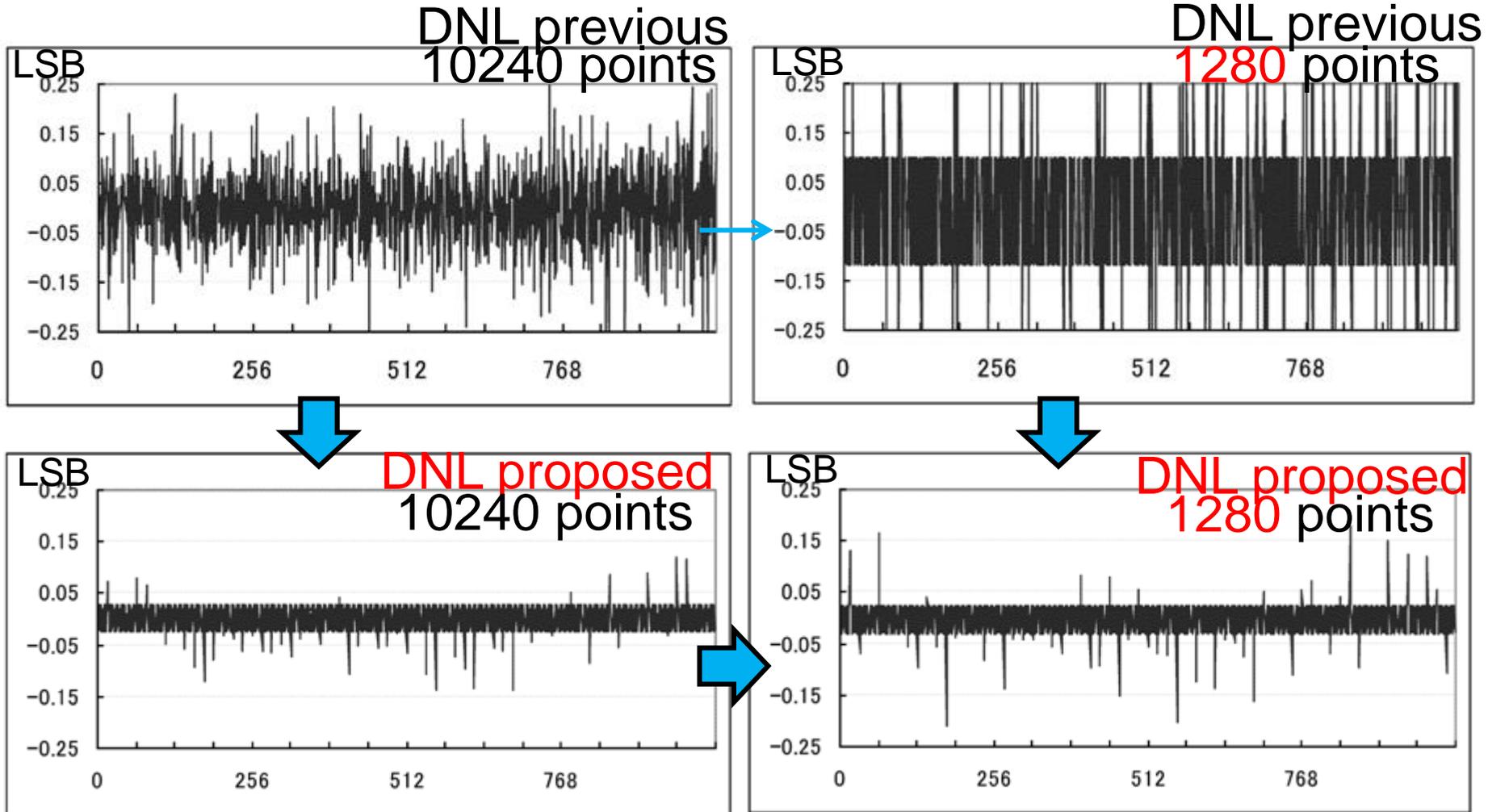
decimated to $1/8$ = 1280 measuring points



Measurement time reduction



Measurement time reduction

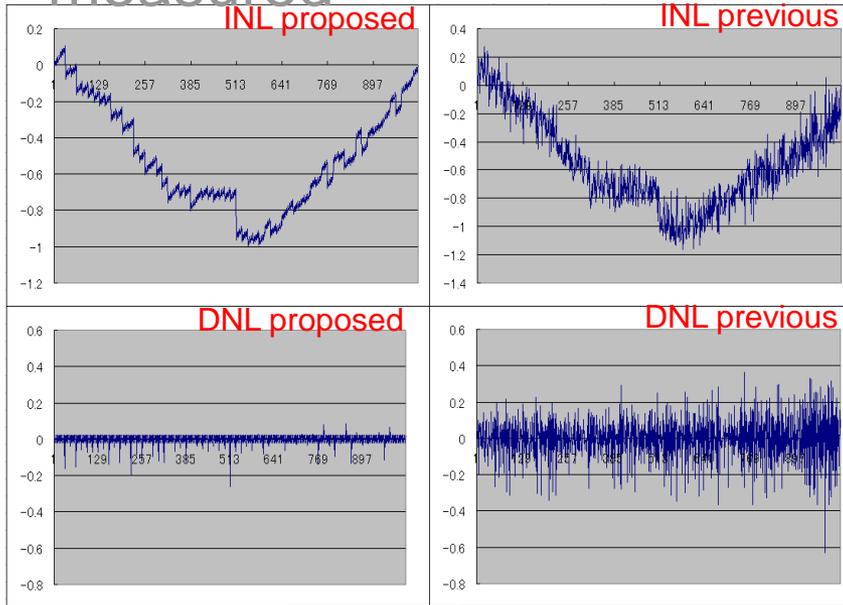


>160,000 equivalent

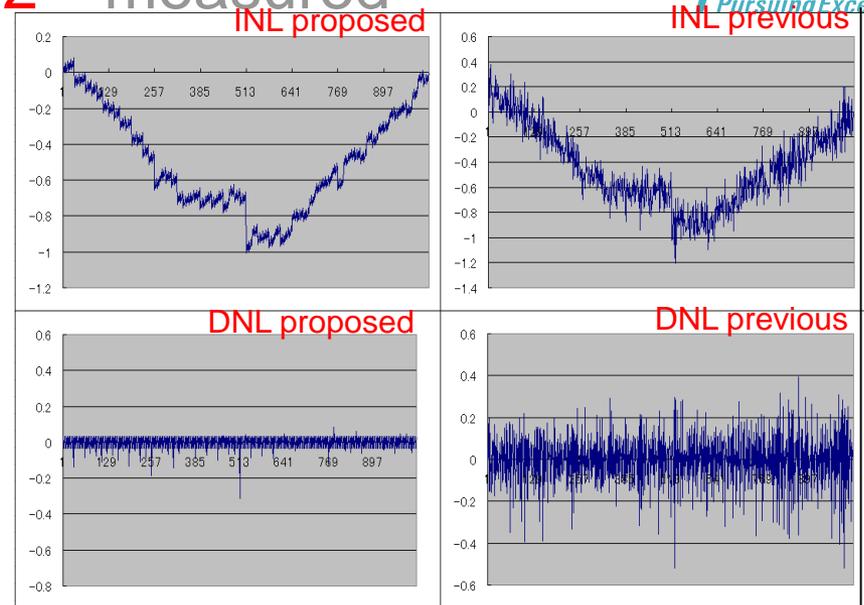
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Experiment results 5

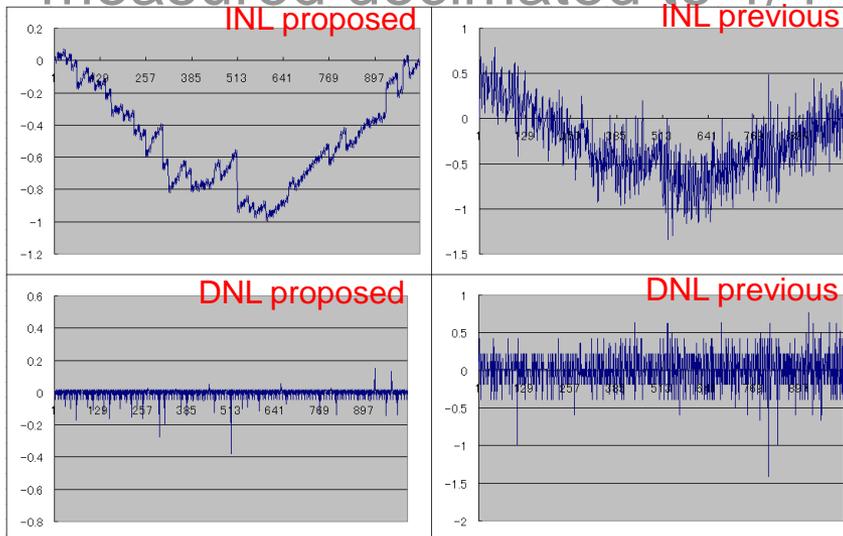
1st measured



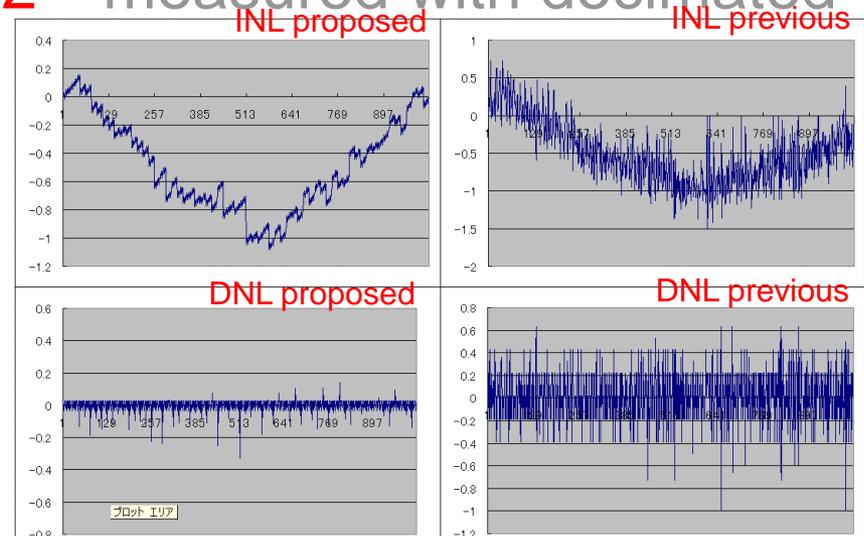
2nd measured



1st measured decimated to 1/4



2nd measured with decimated 1/4



プロットエリア

- **Significant measurement noise reduction to 1/4~1/22**
 - Equivalent x16 (=160,000) measuring point or more
 - Robust data with better repeatability
- **Significant measurement time reduction to 1/8**
 - Better than previous without decimation
 - Robust data with better repeatability
 - **Have realized 1/8 test cost**

- **Proposed novel method to calculate statistically internal physical weights, and to reproduce noise-suppressed ADC**
- **Proposed my method has demonstrated;**
 - 1/4~1/22 measurement noise reduction
 - 1/8 measuring time reduction = 1/8 test cost
- **Has been programed with C, BASIC**
 - evaluation use & LSI tester implementation
 - Binary, thermometer & combined ADC/DAC
 - Now applied in volume production

**I appreciate ex
Renesas Micro Systems Co. Ltd.**